

18-447 Synthesizable Verilog Tutorial

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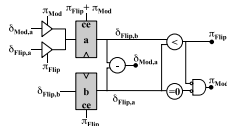
Announcements: none

Handouts: Verilog Tutorial

RTL Verilog Review (The 447 Subset)

What comes to mind
when you think about
hardware designs?

Schematic Capture



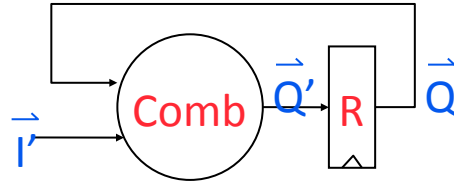
Hardware Description Language

```
always @ (posedge Clk) begin
  if (a >= b) begin
    a <= a - b;
    b <= b;
  end else begin
    a <= b;
    b <= a;
  end
end
```

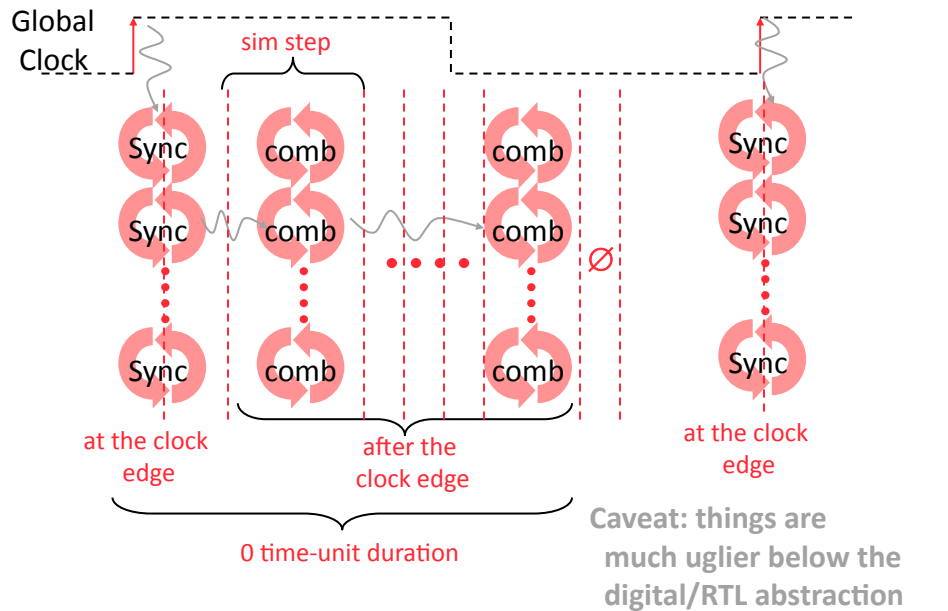
RT-Level Descriptions

- ◆ Textual descriptions of logic circuits
 - different **representation**, but essentially the same **semantics/abstraction** as in schematic capture

So why is it any better?
- ◆ Fine-grain behavior: **synchronous register transfer**
 - a collection of **synchronous** state elements (e.g. registers)
 - a collection of **combinational** logic that computes the next-state (NS or Q') values of state elements based current-state (CS or Q) values of state elements
 - all state elements take on new values simultaneously (implies a globally synchronous clock)



RTL Execution/Timing



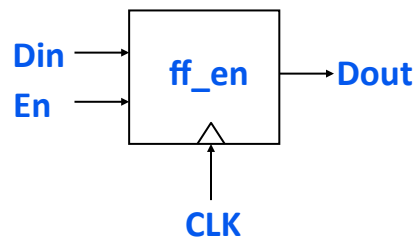
Verilog Modules

- ◆ A module with input and output ports

```
module ff_en( input Din, En, CLK,
              output Dout );
```

```
... body of module ...
```

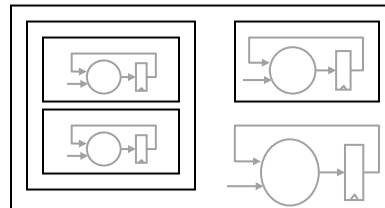
```
endmodule
```



You are NOT writing a Verilog “program”.
You are describing hardware in text

Hierarchical Descriptions

- ◆ Coarse-grain structure
 - A hierarchy of modules
 - A module contains RTL logic and/or, recursively, a network of modules interconnected by wires
 - A module is a **convenience**
 - a design encapsulation/abstraction
 - partitions a design into manageable chunks (vertically and horizontally)
 - reusable design
- ◆ How is a hardware module different from a software function?



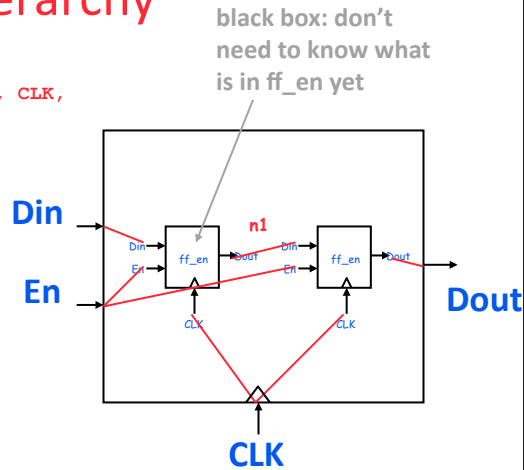
Module Instantiation and Hierarchy

```

module shift_reg (input Din, En, CLK,
                 output Dout );
    wire n1;

    ff_en ff1 (.Din ( Din ),
              .En ( En ),
              .CLK ( CLK ),
              .Dout ( n1 ));

    ff_en ff2 (.Din ( n1 ),
              .En ( En ),
              .CLK ( CLK ),
              .Dout ( Dout ));
endmodule
    
```



Use **named notation** to wire modules:
More likely to get port hookups right!
E.g., connect **ff1's** **Dout** to wire **n1**, ...

Parameters

- ◆ Can parameterize modules:

```

module pmod #(parameter width = 4)
    (input [width-1:0] in,
     output [width-1:0] out);

endmodule
    
```

default parameter value

```

module topmod( ... );
    pmod          m0(.in(a), .out(b));
    pmod #(5)     m1(.in(c), .out(d));
    pmod #(.width(6)) m2(.in(e), .out(f));
endmodule
    
```

width = 4
width = 5
width = 6

Arrays of Modules

- ◆ Hardware frequently has many arrays
 - Verilog distributes **some** connections across arrays for you
 - 1D arrays only
 - Multi-dimensional arrays can be emulated with “hacks”

```

module big_reg #(parameter width = 4)
  (input [width:0] Din,
   input En, CLK,
   output [width:0] Dout );

  wire [width:0]  n1;

  ff_en ff[width:0] ( .Din ( Din ),
                    .En ( En ),
                    .CLK ( CLK ),
                    .Dout ( n1 ));

endmodule
    
```

Arrays of modules just like wires

Automatically distributes
multi-bit and single-bit
inputs (if sizes match)

Combinational Logic

Wires, expressions and assign

```

module ff_en (input Din, En, CLK,
             output Dout);

  wire  cs, ns;

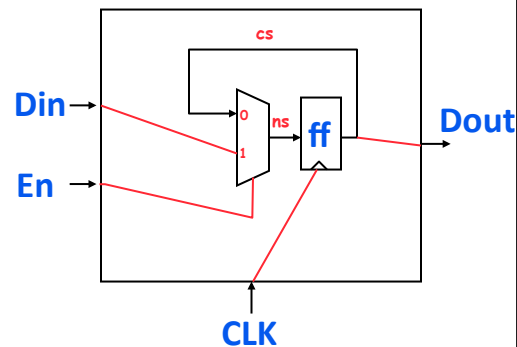
  assign Dout = cs;

  assign ns = En ? Din : cs;

  ff ff1 ( .Din ( ns ),
          .CLK ( CLK ),
          .Dout ( cs ));

endmodule
    
```

continuous assignment:
bind wire to an “expression”



Synchronous Sequential Logic

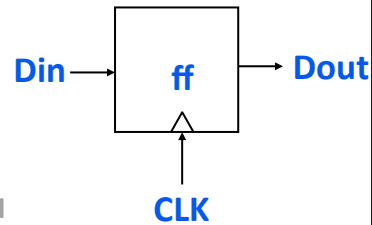
```
module ff (input Din, CLK,
           output Dout);
    reg q;
    assign Dout = q ;
```

“reg” declares a “stateful” variable (as opposed to a stateless wire); reg does not declare a “register”

```
always @ (posedge CLK ) begin
    q <= Din ;
end
endmodule
```

an “always” block (aka process)

A reg variable should only be assigned in *at most one* “always” block!!



Combinational Logic

Procedural Assignment

```
module ff_en (input Din, En, CLK
              output Dout);
```

```
wire cs;
reg ns;
```

```
assign Dout = cs;
```

```
always @ ( Din or cs or En ) begin
    if (En) begin
        ns = Din;
    end else begin
        ns = cs;
    end
end
```

```
ff ff1 ( .Din ( ns ),
        .CLK ( CLK ),
        .Dout ( cs ));
```

```
endmodule
```

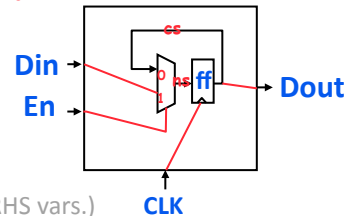
this “reg” is a combinational value

trigger list (must list all RHS vars.) (now, can use always @*)

describing combinational logic as a sequential program (i.e. how to compute ns based on Din or En).

procedure assignment: ns retains the last assigned value at the end of the “always”

A reg variable should appear on LHS in **at most one** always block!! A LHS reg must be assigned **at least once** following **all possible** control paths in a combinational block!!



Combinational Logic

Procedural Assignments Gone Bad

```

module ff_en (input Din, En, CLK,
              output Dout);

  wire  cs;
  reg   ns;

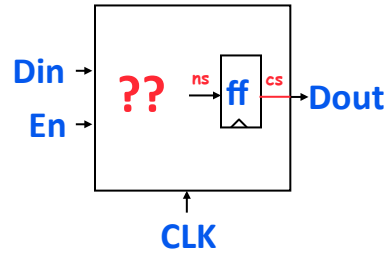
  assign Dout = cs;

  always @ ( Din or cs or En ) begin
    if (En) begin
      ns = Din;
    end // else begin
      // ns = cs;
      // end
  end

  ff ff1 ( .Din ( ns ),
          .CLK ( CLK ),
          .Dout ( cs ));

endmodule

```



What is missing?

Is this still combinational?

What will the synthesis tool do with this?

Combinational Logic

Procedural Assignments Gone Bad **OK??**

```

module ff_en (input Din, En, CLK,
              output Dout);

  wire  cs;
  reg   ns;

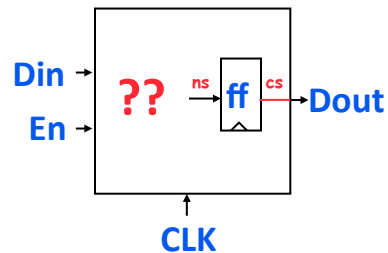
  assign Dout = cs ;

  always @ ( Din or cs or En ) begin
    ns = cs;
    if ( En ) begin
      ns = Din;
    end
  end

  ff ff1 ( .Din ( ns ),
          .CLK ( CLK ),
          .Dout ( cs ));

endmodule

```



If there is control flow, always begin by providing a default assignment to each assigned LHS. Make it a habit.

Combinational Logic

Procedural Assignments Gone Bad

```

module ff_en (input Din, En, CLK,
              output Dout);

  wire  cs;
  reg   ns;

  assign Dout = cs;

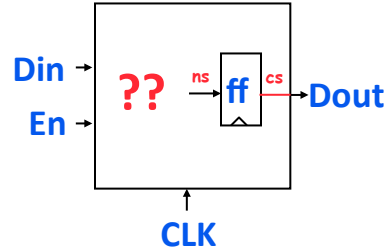
  always @ ( Din or cs /* or En */ )
  begin
    ns = cs ;

    if ( En ) begin
      ns = Din;
    end
  end

  ff ff1 (.Din ( ns ),
         .CLK ( CLK ),
         .Dout ( cs ));

endmodule

```



Incomplete sensitivity lists
considered harmful!

What happens here?

Combinational Logic

Procedural Assignments Gone Bad

```

module ff_en (input Din, En, CLK
              output Dout);

  wire  cs;
  reg   ns;

  assign Dout = cs;

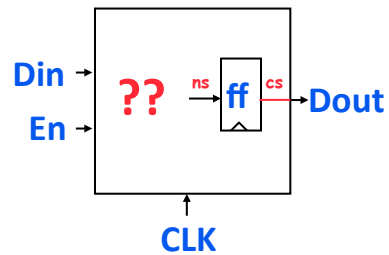
  always @ ( Din or cs or En )
  begin
    ns <= cs;

    if ( En )begin
      ns <= Din;
    end
  end

  ff ff1 (.Din ( ns ),
         .CLK ( CLK ),
         .Dout ( cs ));

endmodule

```



What is the value of ns here?

Combining Combinational and Sequential Logic

```
module ff_en (input Din, En, CLK
              output Dout);

  reg  Dout;      // sequential
  reg  Dout_next; // combinational
```

```
  always @ ( Din or Dout or En ) begin
    if ( En ) begin
      Dout_next = Din;
    end else begin
      Dout_next = Dout;
    end
  end
```

} clearly combinational

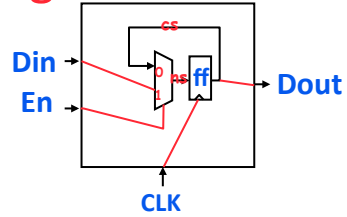
```
  always @ ( posedge CLK ) begin
    Dout <= Dout_next;
  end
```

} clearly synch. sequential

endmodule

Never do

- an always block that has **both** delayed and procedural assigns!!
- a reg assigned to by **both** delayed and procedural assigns!!



Synthesis: under the hood

Inferring Synch Sequential Logic

```
always @ ( posedge CLK ) begin
    LHS <= F( RHS ) ;
end
```

- ◆ Edge triggered “always” block with delayed assignments
 - LHS are reg type and cannot be assigned by another always block
 - RHS are either reg or wire type, can overlap with LHS
 - always block can contain control flow---0 or 1 assign per LHS
- ◆ A register of the appropriate width is instantiated for every LHS reg variable
- ◆ The next-state value on the next clock edge is a combinational function of the RHS variables

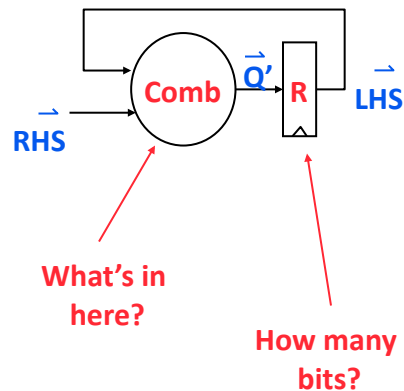
Note: the RHS vars could be combinational or synchronous

Register Inference

```
reg [7:0]    count;
reg         and_bits,
           or_bits,
           xor_bits;

always @ (posedge CLK) begin
    if (reset)
        count <= 0;
    else
        count <= count + 1;

    and_bits <= &count;
    or_bits  <= |count;
    xor_bits <= ^count;
end
```



Register Inference + Retiming

```

reg [7:0]    count;
reg         and_bits,
           or_bits,
           xor_bits;

always @ (posedge CLK) begin
    if (reset)
        count <= 0;
    else
        count <= count + 1;

    and_bits <= &count;
    or_bits <= |count;
    xor_bits <= ^count;
end
    
```



```

reg [7:0]    count;
reg         and_bits,
           or_bits,
           xor_bits;

always @ (posedge CLK) begin
    if (reset)
        count <= 0;
    else
        count <= count + 1;
end

always @ (count) begin
    and_bits = &count;
    or_bits = |count;
    xor_bits = ^count;
end
    
```

Inferring Combinational Logic

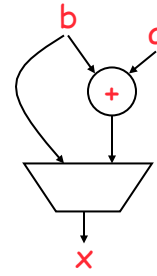
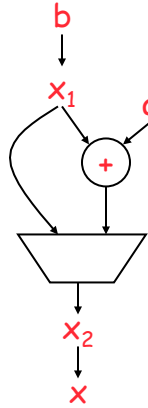
- ◆ Continuous assignments: bind an expression to a wire
 - ◆ Always blocks:
 - procedural assignments to reg-type LHS
 - all input signals are included in the trigger list
 - ⇒ LHS recomputed whenever inputs change
 - { LHS } does not intersect { inputs }
 - ⇒ no feedback cycles
 - all LHS are assigned on all possible control paths
 - ⇒ LHS does not depend on itself ⇒ no storage
- ⇒ A combinational relationship between RHS and LHS

Procedural Descriptions

```
{
  x = b;
  if (y)
    x = x + a;
}
```

```
{
  x1 = b;
  if (y)
    x2 = x1 + a;
  else
    x2 = x1;
  . . . . .
  x = xmax;
}
```

```
{
  if (y)
    x = b + a;
  else
    x = b;
}
```



to flatten out a procedural description, convert to single-assignment

Sequentialized Description

```
output [15:0] Z;
reg [15:0] Z;

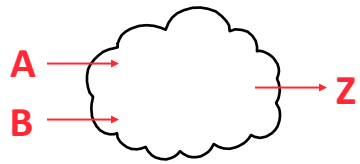
input [15:0] A;
input [15:0] B;

integer i;

always @ (A or B) begin
  Z=0;

  for(i=0;i<=15;i=i+1) begin
    if (A[i]) begin
      Z=Z+(B<<i);
    end
  end
end

end
```



Try writing *that* in a continuous assignment expression

Synthesizable Loops

- ◆ Static loops are allowed
 - good for scanning across fixed size vectors and arrays
- ◆ A loop must be statically unrollable, i.e.,
 - loop index must be integer type
 - loop index initial value must statically resolve to a constant
 - valid loop index operations are +, -
 - the valid loop condition test must test against a static limit using relational operators: <, <=, >, >=

Some tools even support bounded-recursive functions. We do not suggest this in 18-447.

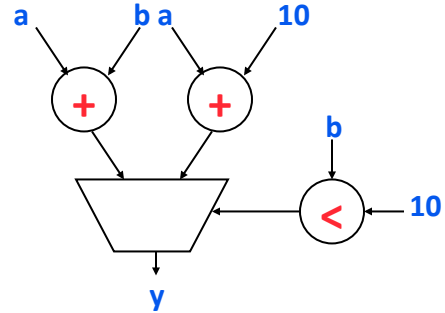
Combinational Logic

- ◆ In general
 - RTL Synthesis tools let you control the combinational logic by the structure of your expressions
 - With optimizations turned off, tools translate an expression literally and faithfully
 - Most optimizations are turned-off by default
 - RTL Synthesis tools can infer “custom-width” {adders, subtractors, comparators}, multipliers and multiplexers, everything else is mapped as random logic
 - Simple syntax-directed translation
 - You can instantiate library modules explicitly to improve performance
 - Tool- and target-specific “pragmas” or “directives” permit fine-tuning.

Simple Example

◆ The statement

```
if (b < 10)
    y = a + b;
else
    y = a + 10;
```



◆ Maps to

- A comparator that compares the value of **b** with **10**
- An adder that has **a** and **b** as inputs
- An adder that has **a** and **10** as inputs
- A multiplexer (implied by the `if` statement) that controls the final value of **y**

◆ The width of the datapath depends type of **a**, **b** and **y**

Case-Switch

```
output [7:0] out;
input [2:0] in;
reg [7:0] out;

always @( in ) begin
    case (in)
        3'h0: out=1;
        3'h1: out=2;
        3'h2: out=4;
        3'h3: out=8;
        3'h4: out=16;
        3'h5: out=32;
        default: out=x;
    endcase
end
```

Watch out: keywords
different from C/C++

Tells synthesis that
everything else is
don't care!

Verilog Compiler Directives

- ◆ Insert external file
 - ``include "filename.v"`
 - inserts `filename.v` at that line
 - ◆ Macros
 - ``define MACRONAME value`
 - Replace ``MACRONAME` with value everywhere in code
 - ◆ Macro-controlled conditionals
 - Use different code depending on which macros are defined
 - ``ifdef SYNTHESIZE`
 - ``include "something_synthesizable.v"`
 - ``else`
 - ``include "something_not_synthesizable.v"`
 - ``endif`
 - Also: ``ifndef` "if not defined"
- Don't forget
the backticks!**

Default Net Type

- ◆ By default, most tools will assume that undeclared values are 1-bit wires
- ◆ Instead, you can turn this "feature" off!

Before	After
<pre style="font-family: monospace; font-size: 0.9em;"> module test (input [3:0] a, output [3:0] b); wire [3:0] typo; assign typo = a; assign b = typo; endmodule </pre> <p style="font-size: 0.8em; margin-top: 10px;">Assumes "typo" is a one-bit wire (and doesn't tell you!)</p>	<pre style="font-family: monospace; font-size: 0.9em;"> `default_nettype none module test (input wire [3:0] a, output wire [3:0] b); wire [3:0] typo; assign typo = a; assign b = typo; endmodule </pre> <p style="font-size: 0.8em; margin-top: 10px;">Gives error: "typo" undeclared</p>
	<p style="font-size: 0.8em;">Now need to explicitly declare inputs and outputs as "wires"</p>

Readings

- ◆ Highly recommended readings,
 - HDL Compiler for Verilog: Reference Manual, Synopsys Documentation
 - Guide to HDL Coding Styles for Synthesis, Synopsys Documentation